

University of Moratuwa, Sri Lanka
B. Sc. Engineering Honours Degree Course, Semester 2 Examination
EE2092 – THEORY OF ELECTRICITY

Short Answers

Question 1

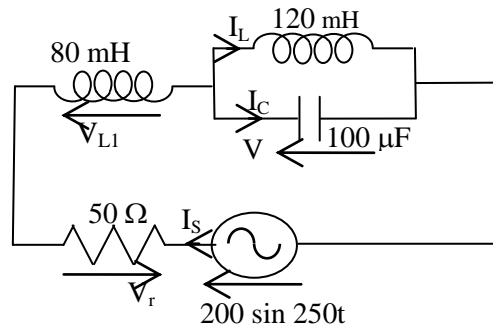


Figure Q1

- (a) Write down the differential equations governing the behaviour of the circuit.

$$200 \sin 250t = 50 i_s + 0.08 \frac{di_s}{dt} + v$$

$$i_s = i_L + i_C$$

$$v = 0.120 \frac{di_C}{dt} = 10^4 f i_C dt$$

[2 marks]

- (b) Calculate the total impedance of the circuit in ohm.

$$Z = 50 + j 0.08 \times 250 + j 0.12 \times 250 // (10^4 / j 250) = 50 + j 20 + \frac{j 30 \times (-j 40)}{j 30 - j 40}$$

$$= 50 + j 140 \Omega \text{ (or } 148.66 \angle 70.35^\circ \Omega)$$

[2 marks]

- (c) Calculate the current I_s supplied from the source.

$$I_s = \frac{\frac{200}{\sqrt{2}} \angle 0^\circ}{148.66 \angle 70.35^\circ} = 0.951 \angle -70.35^\circ \text{ A}$$

[1 mark]

- (d) Determine the currents I_L and I_C .

$$I_L = \frac{-j 40}{j 30 - j 40} I_s = 4 I_s = 3.805 \angle -70.35^\circ \text{ A}$$

$$I_C = \frac{j 30}{j 30 - j 40} I_s = -3 I_s = 2.854 \angle 109.65^\circ \text{ A}$$

[2 marks]

- (e) Determine the voltages V_r , V_{L1} and V .

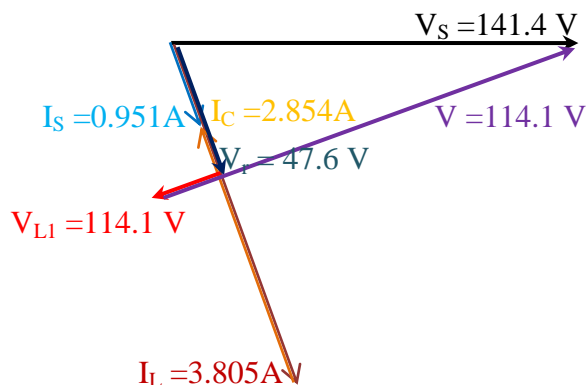
$$V_r = 50 I_s = 47.57 \angle -70.35^\circ \text{ V}$$

$$V_{L1} = j 250 \times 0.08 I_s = 19.02 \angle 19.65^\circ \text{ V}$$

$$V = j 30 \times I_L = 114.1 \angle -70.35^\circ \text{ V}$$

[2 marks]

- (f) Sketch a phasor diagram showing all the voltages and currents in the circuit.



[3 marks]

- (g) If the supply frequency is allowed to vary, determine the frequency at which series resonance would occur.

$$j\omega \times 0.08 + \frac{j0.12\omega \times \left(\frac{10^4}{j\omega}\right)}{j0.12\omega + \frac{10^4}{j\omega}} = 0$$

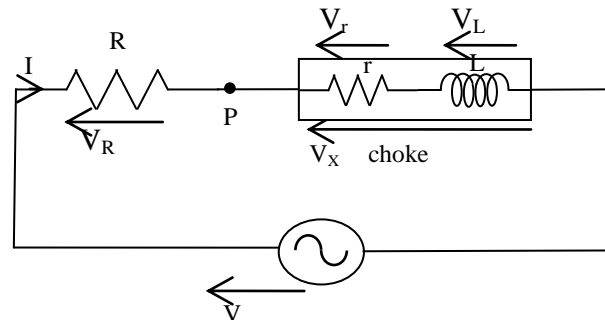
$$j\omega 0.08 \times j\omega 0.12 + 10^4 \times 0.08 + 0.12 \times 10^4 = 0$$

$$0.0096 \omega^2 = 0.2 \times 10^4$$

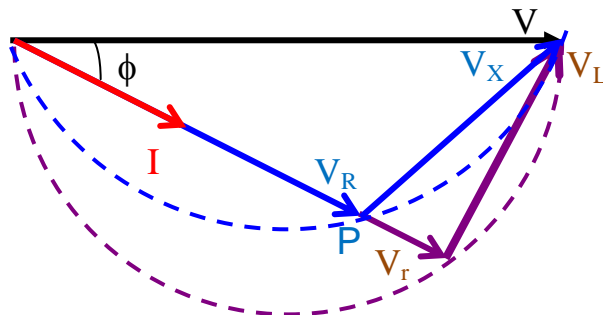
$$\omega = 456.4 \text{ rad/s (or 72.64 Hz)}$$

[2 marks]

Question 2



- (a) A practical choke (inductance L and series resistance r), connected in series with a pure resistance R , is supplied from a constant voltage source (V, ω). If the resistance R is allowed to vary, sketch the locus of the voltage of the intersection P between the choke and the resistor.



[3 marks]

- (b) One of the readings in Q2(a) corresponds to $V = 250\text{V}$ at 50 Hz , current $I = 10\text{A}$ lagging voltage by 30° , the voltage magnitudes across the resistor $V_R = 150\text{V}$ and choke $V_X = 141.6\text{V}$. Complete the phasor diagram and calculate the inductance L and the resistance r of the choke.

Phasor diagram is completed as in above diagram

$$V_R + V_r = V \cos \phi = 250 \cos 30^\circ = 216.5 \text{ V}$$

$$V_R = 150 \text{ V}, \therefore V_r = 216.5 - 150 = 66.5 \text{ V}$$

$$r = 66.5/10 = 6.65 \Omega$$

$$V_L = V \sin \phi = 250 \sin 30^\circ = 125 \text{ V}$$

$$L = 125/(10 \times 2\pi \times 50) = 39.79 \text{ mH}$$

[3 marks]

(c) Derive the non-coupled equivalent circuit of a two winding transformer.

Consider the coupled circuit shown

From Kirchoff's voltage law

$$V_p = L_p p i_p - M p i_s$$

and

$$V_s = -(L_s p i_s - M p i_p)$$

For near ideal transformers

$$|V_p| : |V_s| \approx a$$

$$|i_p| : |i_s| \approx 1/a$$

Thus

V_p and aV_s are comparable quantities

i_p and i_s/a are comparable quantities.

Using these comparable variables

$$V_p = L_p p i_p - aM p i_s/a$$

and

$$aV_s = -(a^2 L_s p i_s/a - aM p i_p)$$

If i_p and i_s/a are unequal,

Leakage current = $i_p - i_s/a$

re-formulated in terms of this leakage current as

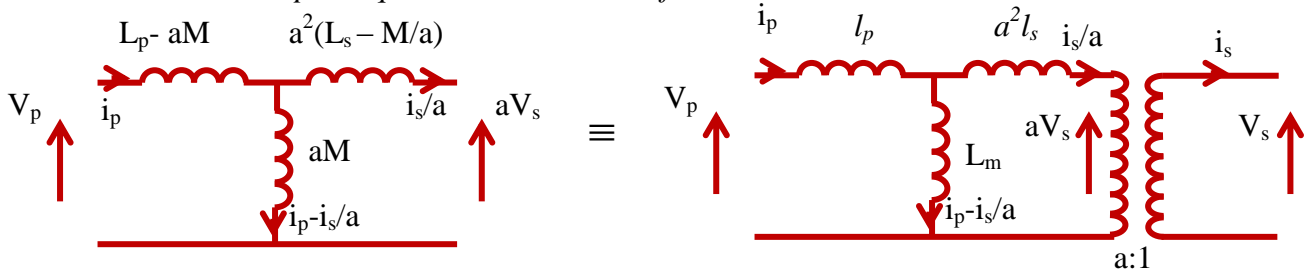
$$\begin{aligned} V_p &= L_p p i_p - aM p i_s/a + aM p i_p - aM p i_p \\ &= (L_p - aM) p i_p + aM p .(i_p - i_s/a) \end{aligned}$$

and

$$\begin{aligned} aV_s &= -(a^2 L_s p i_s/a - aM p i_p) + aM p i_s/a - aM p i_s/a \\ &= -a^2 (L_s - \frac{M}{a}) p i_s/a + aM p .(i_p - i_s/a) \end{aligned}$$

Equations are expressed either in terms of i_p and $i_p - i_s/a$ or in terms of i_s/a and $i_p - i_s/a$.

Allows a non-coupled equivalent circuit to be formed



[2 marks]

A 80/20kV transformer has a primary winding inductance of 16H, and a coupling coefficient of 0.95. Using any standard equations derive the values of the inductive impedances of all the elements of the non-coupled equivalent circuit.

$$a = 80/20 = 4$$

$$L_p = 16 \text{ H}, L_s = L_p/a^2 = 16/4^2 = 1 \text{ H}$$

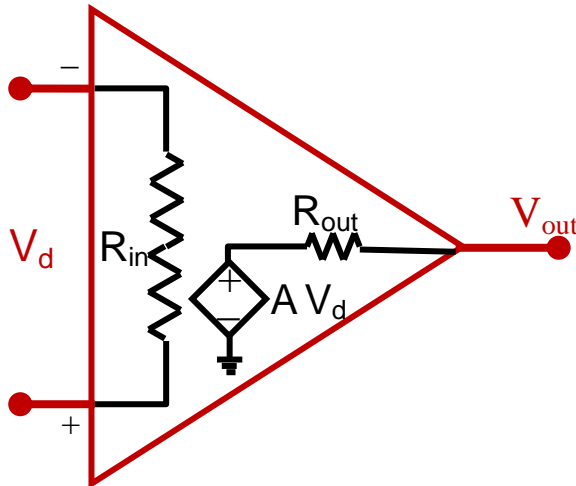
$$M = k \sqrt{L_p L_s} = 0.95 \times \sqrt{16 \times 1} = 3.8 \text{ H}$$

$$l_p = L_p - a M = 16 - 4 \times 3.8 = 0.8 \text{ H}$$

$$l_s = L_s - M/a = 1 - 3.8/4 = 0.05 \text{ H}$$

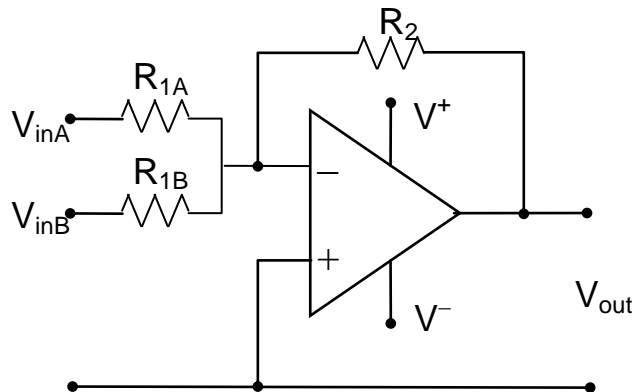
[2 marks]

- (d) Sketch the equivalent circuit of an operational amplifier and show that the voltage difference of the input terminals may be taken as a virtual zero.



Since the gain A of an operational amplifier is very large (10^5 to 10^8), for any finite V_{out} , V_d becomes extremely small, and when the lower terminal of the input is earthed, the upper terminal becomes a virtual earth for all practical purposes. **[2 marks]**

- (e) It is required to construct a summing amplifier to add 75% of input A voltage and 25% of input B voltage. If the resistance connected across the operational amplifier is $1\text{ k}\Omega$, use a diagram to show what should be the input resistances to be connected for the two inputs?



$$V_{out} = -R_2 \left(\frac{V_{inA}}{R_{1A}} + \frac{V_{inB}}{R_{1B}} \right)$$

$$= - (0.75 V_{inA} + 0.25 V_{inA})$$

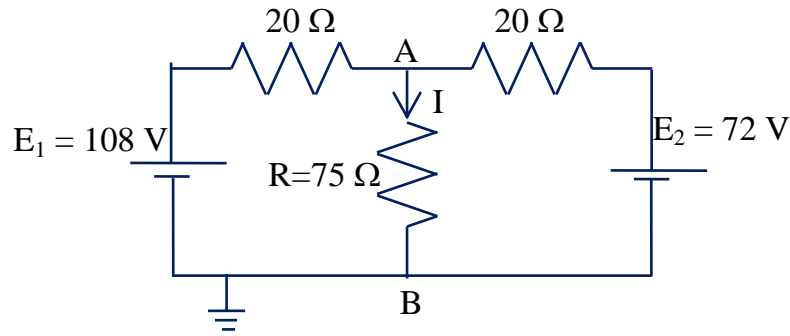
$$R_2 = 1\text{ k}\Omega$$

$$\therefore R_2/R_{1A} = 0.75, \text{ and } R_2/R_{1B} = 0.25,$$

$$\text{i.e. } R_{1A} = 1000/0.75 = 1.33\text{ k}\Omega \text{ and } R_{1B} = 1000/0.25 = 4\text{ k}\Omega$$

[2 marks]

Question 3



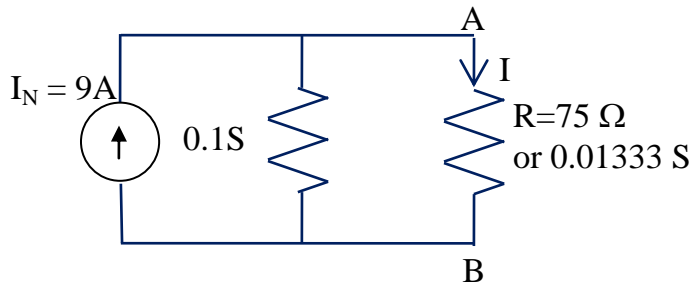
- (a) For the circuit shown in figure Q3ab, obtain the Norton's equivalent circuit across AB and hence obtain the current I.

Short-circuiting AB, it is clearly seen that the short circuit current across AB is

$$I_{SCAB} = 108/20 + 72/20 = 9 \text{ A, i.e. } I_N = 9 \text{ A}$$

Equivalent Norton's admittance across AB is the parallel of 20Ω and 20Ω.

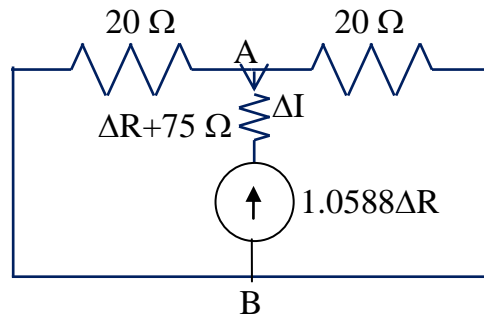
$$\text{i.e. } Y_N = 0.05 + 0.05 = 0.10 \text{ S}$$



$$\text{Current } I = 9 \times 0.013333 / (0.1 + 0.01333) = 1.0588 \text{ A}$$

[3 marks]

- (b) Using compensation theorem, determine the resistance to which R must be changed if the desired current I in figure Q3ab is 1.0A.



Using compensation theorem, required change in current = $1 - 1.0588 = -0.0588 \text{ A}$

Using this together with the circuit,

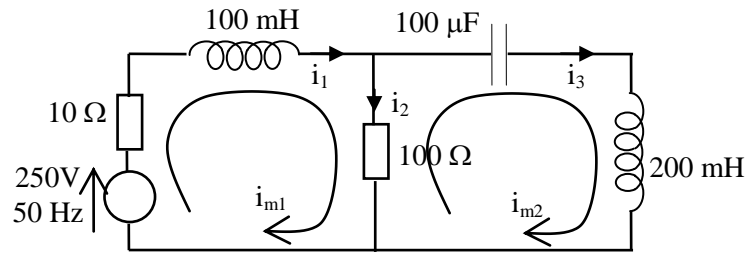
$$\Delta I = -0.0588 \text{ A} = \frac{1.0588 \times \Delta R}{75 + \Delta R + 10} 1.0588$$

$$\text{Giving } \Delta R = 85 \times 0.0588 = 5.0 \text{ } \Omega$$

$$\therefore \text{required value of } R = 75 + 5 = 80 \text{ } \Omega$$

[2 marks]

- (c) For the circuit shown in figure Q3c, write down the branch-mesh incidence matrix and the branch impedance matrix.



$$100 \text{ mH} \rightarrow 100 \times 10^{-3} \times 100\pi = j 31.416 \Omega$$

$$200 \text{ mH} \rightarrow 200 \times 10^{-3} \times 100\pi = j 62.832 \Omega$$

$$100 \mu\text{F} \rightarrow 1/(j100 \times 10^{-6} \times 100\pi) = -j31.831 \Omega$$

$$\text{Branch mesh incidence matrix} = \begin{bmatrix} 1 & 0 \\ 1 & -1 \\ 0 & 1 \end{bmatrix}$$

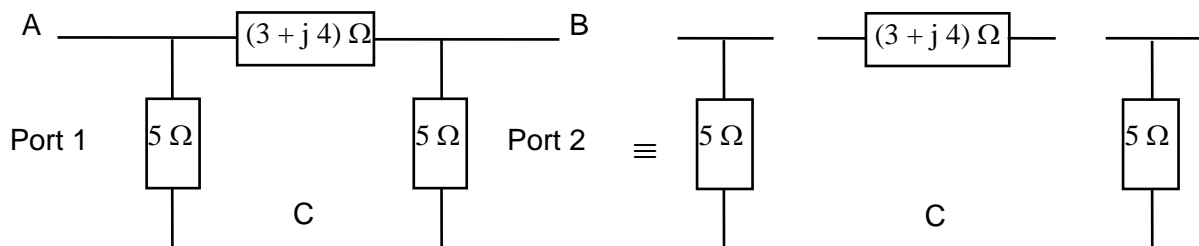
$$\text{Branch impedance matrix} = \begin{bmatrix} 10 + j31.416 & 0 & 0 \\ 0 & 100 & 0 \\ 0 & 0 & j62.832 - j31.831 \end{bmatrix} \quad [3 \text{ marks}]$$

Hence determine the mesh impedance matrix.

$$[Z_m] = \begin{bmatrix} 1 & 1 & 0 \\ 0 & -1 & 1 \end{bmatrix} \begin{bmatrix} 10 + j31.416 & 0 & 0 \\ 0 & 100 & 0 \\ 0 & 0 & j62.832 - j31.831 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ 1 & -1 \\ 0 & 1 \end{bmatrix}$$

$$= \begin{bmatrix} 110 + j31.416 & -100 \\ -100 & 100 + j31.00 \end{bmatrix} \quad [2 \text{ marks}]$$

- (d) Determine the ABCD matrix of the two-port network shown in figure Q3de.

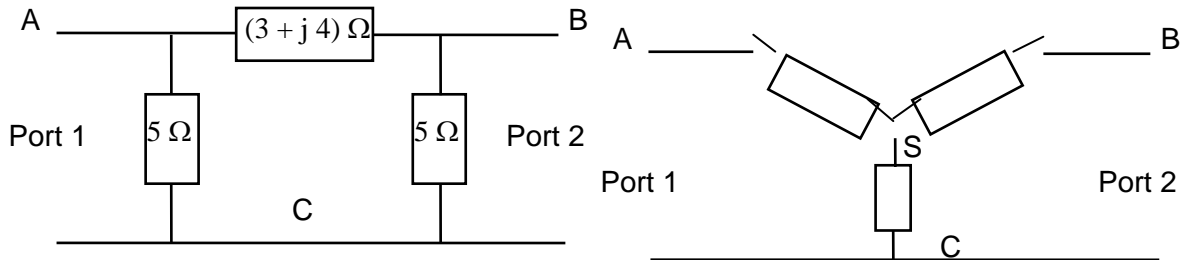


ABCD matrix can be obtained by the product of the individual ABCD matrices

$$\begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0.2 & 1 \end{bmatrix} \begin{bmatrix} 1 & 3 + j4 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ 0.2 & 1 \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0.2 & 1 \end{bmatrix} \begin{bmatrix} 1.6 + j0.8 & 3 + j4 \\ 0.2 & 1 \end{bmatrix}$$

$$\text{i.e. } \begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} 1.6 + j0.8 & 3 + j4 \\ 0.52 + j0.16 & 1.6 + j0.8 \end{bmatrix} \quad [2 \text{ marks}]$$

- (e) Convert the circuit shown in figure Q3de to an equivalent star-connected network between nodes A, B and C and redraw the circuit.



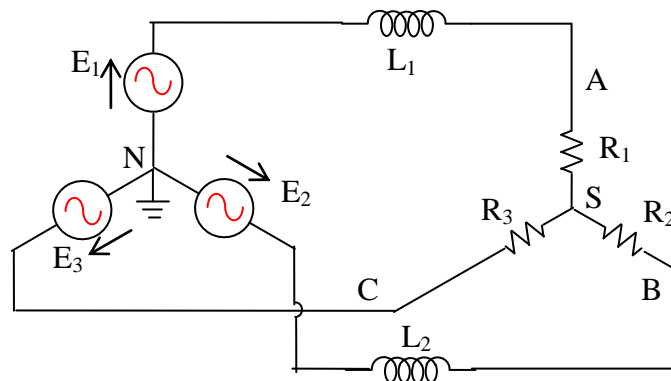
$$Z_{AS} = Z_{AB} \times Z_{AC} / (Z_{AB} + Z_{BC} + Z_{CA}) = (3 + j4) \times 5 / (3 + j4 + 5 + 5) = 1.838 \angle 36.0^\circ \Omega$$

$$Z_{BS} = Z_{BA} \times Z_{BC} / (Z_{AB} + Z_{BC} + Z_{CA}) = (3 + j4) \times 5 / (3 + j4 + 5 + 5) = 1.838 \angle 36.0^\circ \Omega$$

$$Z_{CS} = Z_{CB} \times Z_{CA} / (Z_{AB} + Z_{BC} + Z_{CA}) = 5 \times 5 / (3 + j4 + 5 + 5) = 1.838 \angle 17.1^\circ \Omega \quad [2 \text{ marks}]$$

Question 4

- (a) The supply shown in figure Q4(a) is a balanced 3phase, 400V, 50 Hz source. If $L_1 = L_2 = 100 \text{ mH}$, and $R_1 = R_2 = R_3 = 100 \Omega$, determine the voltage of the star point S.



$$L_1 = L_2 = 100 \text{ mH} \rightarrow 100 \times 10^{-3} \times 100\pi = j 31.416 \Omega$$

$$V_S = V_{SN} = \frac{\sum Y \cdot V}{\sum Y} = \frac{\frac{230 \angle 0^\circ}{100 + j31.416} + \frac{230 \angle -120^\circ}{100} + \frac{230 \angle 120^\circ}{100 + j31.416}}{\frac{1}{100 + j31.416} + \frac{1}{100} + \frac{1}{100 + j31.416}}$$

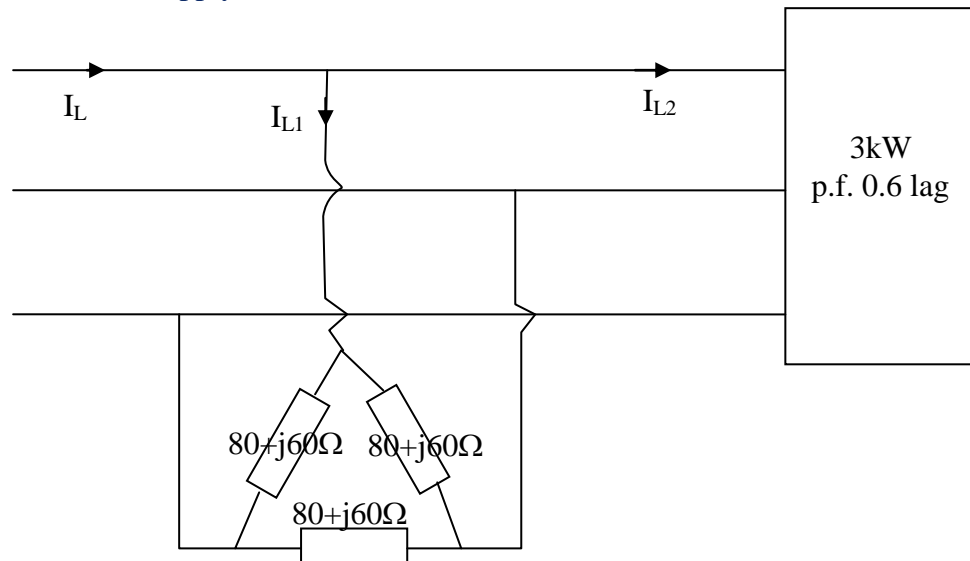
$$= 23.95 \angle -155.98^\circ \text{ V}$$

$$I_A = \frac{230 \angle 0^\circ - 23.95 \angle -155.98^\circ}{100 + j31.416} = 2.41 \angle -15.22^\circ \text{ V} \quad [4 \text{ marks}]$$

Hence determine the potential at A

$$V_A = V_S + 100 I_A = 23.95 \angle -155.98^\circ + 100 \times 2.41 \angle -15.22^\circ = 223.3 \angle -19.1^\circ \text{ V} \quad [2 \text{ marks}]$$

- (b) A balanced 400 V, 3-phase, 50 Hz supply feeds (i) a 3-phase delta-connected balanced load consisting of arms of value $(80+j60)\ \Omega$ each and (ii) a three phase motor load of 3 kW at a power factor of 0.6 lag, Determine the line current, power factor and the active power at the supply.



$$I_{L1} = \sqrt{3} \times \frac{400}{80 + j60} = 6.928 \angle -36.9^\circ \text{ A}$$

$$I_{L2} = \frac{3000}{\sqrt{3} \times 400 \times 0.6} = 7.217 \angle -53.1^\circ \text{ A}$$

$$I_L = I_{L1} + I_{L2} = 6.928 \angle -36.9^\circ + 7.217 \angle -53.1^\circ = 14.0 \angle -45.16^\circ$$

$$\text{Power factor} = 0.705 \text{ lag}$$

$$\text{Active Power} = \sqrt{3} \times 400 \times 14.0 \times 0.705 = 6839 \text{ W}$$

[3 marks]

Determine also the rating of the delta connected capacitor bank required to improve the overall power factor to 0.95 lagging.

$$\text{Reactive Power} = \sqrt{3} \times 400 \times 14.0 \times \sin 45.15^\circ = 6827 \text{ var}$$

$$\text{New power factor} = 0.95 \rightarrow 18.19^\circ$$

$$\text{New reactive Power} = 6839 \times \tan 18.19^\circ = 2248 \text{ var}$$

$$\text{Rating of capacitor bank} = 6827 - 2248 = 4579 \text{ var}$$

$$\text{Rating of each capacitor} = 4579/3 = 1526 \text{ var}$$

[2 marks]

$$[or 100 \times \pi \times C \times 400^2 = 1526 \rightarrow 30.36 \mu\text{F} - \text{not necessary}]$$

- (c) The phase components of the currents in an unbalanced system A-B-C is given as $I_A = 20 \angle -30^\circ \text{ A}$, $I_B = 15 \angle 60^\circ \text{ A}$ and $I_C = 20 \angle -120^\circ \text{ A}$. Determine the sequence components of the currents.

$$\begin{bmatrix} I_{A0} \\ I_{A1} \\ I_{A2} \end{bmatrix} = \frac{1}{3} \begin{bmatrix} 1 & 1 & 1 \\ 1 & \alpha & \alpha^2 \\ 1 & \alpha^2 & \alpha \end{bmatrix} \begin{bmatrix} 20 \angle -30^\circ \\ 15 \angle 60^\circ \\ 20 \angle -120^\circ \end{bmatrix}$$

$$I_{A0} = 1/3(20 \angle -30^\circ + 15 \angle 60^\circ + 20 \angle -120^\circ) = 6.87 \angle -44.03^\circ \text{ A}$$

$$I_{A1} = 1/3(20 \angle -30^\circ + 15 \angle 180^\circ + 20 \angle 120^\circ) = 3.54 \angle 136.4^\circ \text{ A}$$

$$I_{A2} = 1/3(20 \angle -30^\circ + 15 \angle 300^\circ + 20 \angle 0^\circ) = 16.79 \angle -27.16^\circ \text{ A}$$

[2 marks]

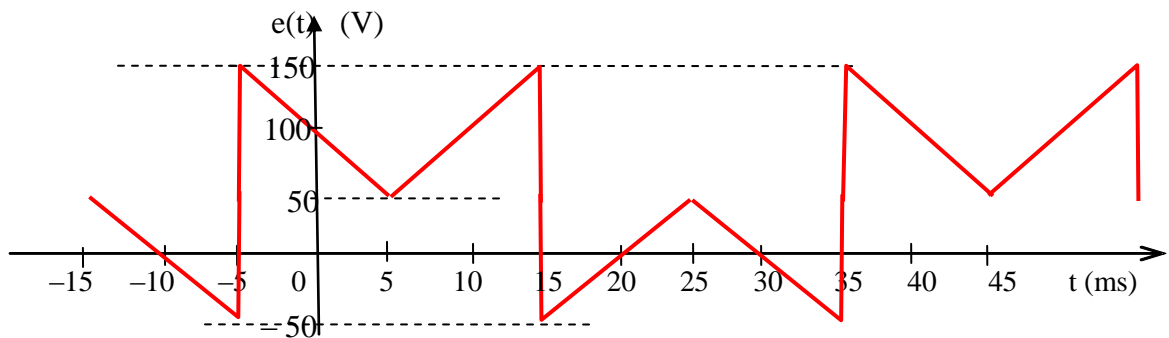
If the corresponding sequence components of the voltages are $V_{A0} = 50\angle-30^\circ$ V, $V_{A1} = 200\angle 0^\circ$ V, $V_{A2} = 50\angle-30^\circ$ V, determine the total active power. [1 mark]

$$P = 3 V_{A0} I_{A0} \cos \phi_0 + 3 V_{A1} I_{A1} \cos \phi_1 + 3 V_{A2} I_{A2} \cos \phi_2$$

$$P = 3(50 \times 6.87 \cos 14.03^\circ + 200 \times 3.54 \cos 136.4^\circ + 50 \times 16.79 \cos 2.84^\circ) = 1979 \text{ W}$$

Question 5

- (a) For the voltage waveform $e(t)$ shown in figure Q5a, write down any simplifications that can be used to determine the Fourier series and determine the first 4 significant terms of the Fourier series.

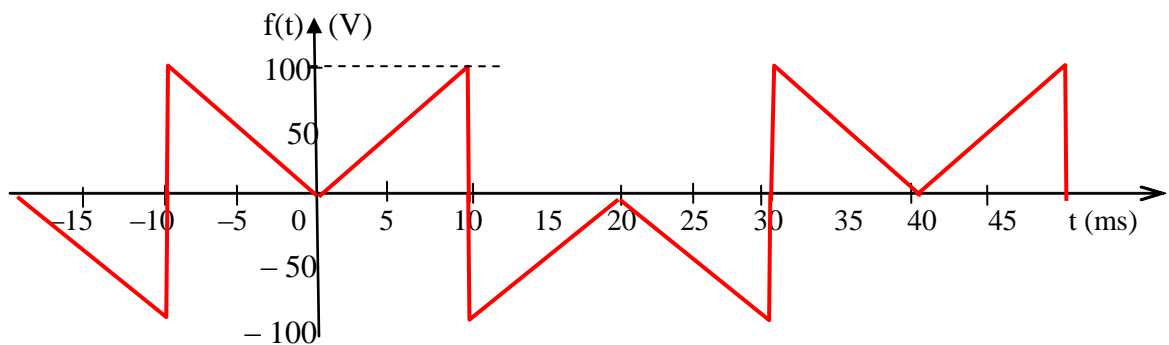


Shift waveform downwards by 50 V and left by 5ms (becomes even and gives a cosine series) [or right by 5ms - becomes odd and gives a sine series]

New waveform $f(t)$, period $T = 40$ ms

$$F(t) = e(t - 0.005) - 50$$

[2 marks]



Waveform $f(t)$ has even symmetry and half-wave symmetry with mean value zero.

Thus for $f(t)$, the Fourier coefficients are given as

$$A_0 = 0 \text{ (since mean value} = 0), B_n = 0 \text{ (since even function),}$$

$$A_n = 0 \text{ for even } n \text{ (since half wave symmetry)}$$

$$f(t) = 10^4 t \quad 0 \leq t \leq 0.01$$

Thus for odd n ,

$$A_n = 4 \times \frac{2}{T} \int_0^{\frac{T}{4}} f(t) \cdot \cos n\omega_0 t dt = \frac{8}{0.04} \int_0^{0.01} 10^4 t \cdot \cos n50\pi t \cdot dt$$

$$A_n = 2 \times 10^6 \left[\frac{t \cdot \sin n50\pi}{n \cdot 50\pi} \Big|_0^{0.01} - \int_0^{0.01} \frac{\sin n50\pi}{n \cdot 50\pi} \right] = 40 \times 10^3 \left[\frac{0.01 \times \sin \frac{n\pi}{2}}{n \cdot \pi} + \frac{\cos n50\pi}{n \cdot \pi} \Big|_0^{0.01} \right]$$

$$= \frac{40 \times 10^3}{n\pi} \left[0.01 \times \sin \frac{n\pi}{2} - \frac{1}{50n\pi} \right] = \frac{400}{n\pi} \sin \frac{n\pi}{2} - \frac{800}{(n\pi)^2} \quad \text{for odd } n$$

[3 marks]

$$A_1 = \frac{400}{\pi} \sin \frac{\pi}{2} - \frac{800}{\pi^2} = \frac{400}{\pi} - \frac{800}{\pi^2} = 46.27$$

$$A_3 = \frac{400}{3\pi} \sin \frac{3\pi}{2} - \frac{800}{9\pi^2} = -\frac{400}{3\pi} - \frac{800}{9\pi^2} = -51.44$$

$$A_5 = \frac{400}{5\pi} \sin \frac{5\pi}{2} - \frac{800}{25\pi^2} = \frac{80}{\pi} - \frac{32}{\pi^2} = 22.22$$

$$A_7 = \frac{400}{7\pi} \sin \frac{7\pi}{2} - \frac{800}{49\pi^2} = -\frac{400}{7\pi} - \frac{800}{49\pi^2} = -19.84$$

[1 mark]

Thus the Fourier series of $f(t)$ is

$$f(t) = 46.27 \cos 50\pi t - 51.44 \cos 150\pi t + 22.22 \cos 250\pi t - 19.84 \cos 350\pi t + \dots$$

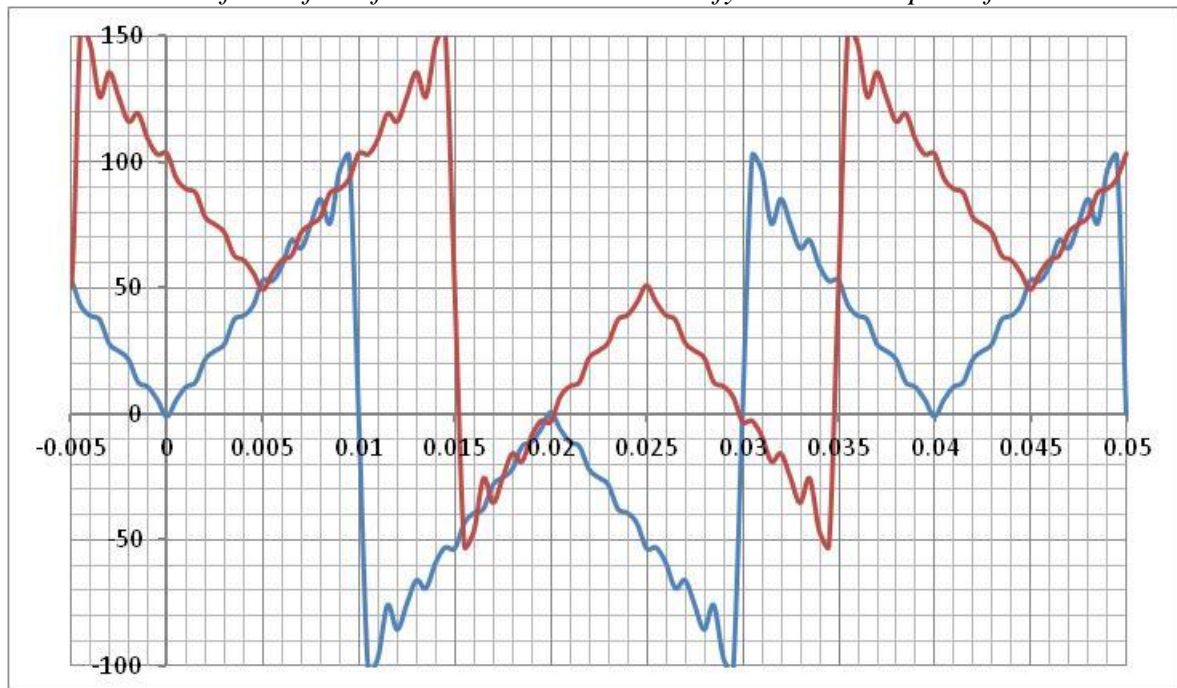
and the Fourier series of the original waveform is

$$e(t) = 50 + e(t+0.005)$$

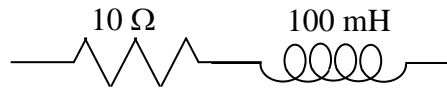
$$e(t) = 50 + 46.27 \cos 50\pi(t+0.005) - 51.44 \cos 150\pi(t+0.005) + 22.22 \cos 250\pi(t+0.005) - 19.84 \cos 350\pi(t+0.005) + \dots$$

[1 mark]

Reconstruction of waveform from Fourier series to verify answer. Not part of answer.



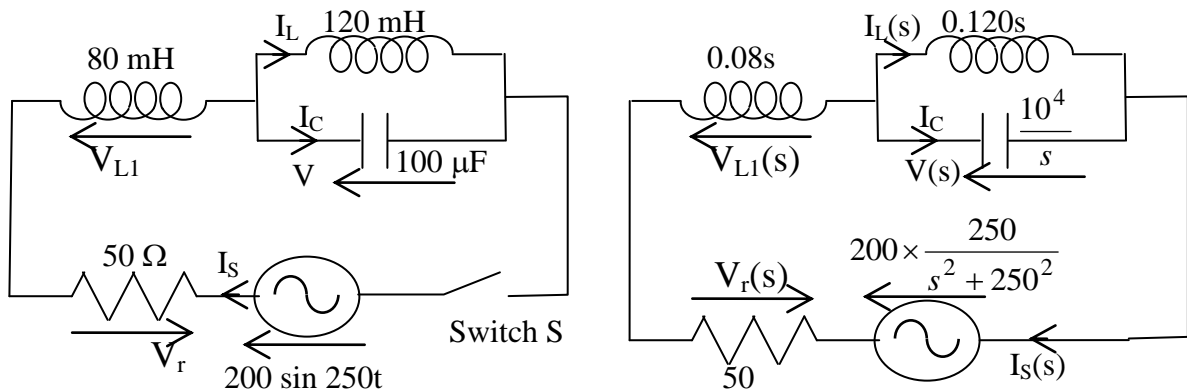
- (b) If a current $i(t) = 1 + 10 \sin 100t + 3 \sin (300t + \pi/3) + 2 \sin(500t - \pi/6)$ is passed through a series combination of a resistor $R = 10 \Omega$ and an inductor $L = 100 \text{ mH}$, determine the Fourier series of the resulting voltage $v(t)$.



$$\begin{aligned}
 V(t) &= R.i(t) + L.di(t)/dt \\
 &= 10 \times (1 + 10 \sin 100t + 3 \sin (300t + \pi/3) + 2 \sin(500t - \pi/6)) \\
 &\quad + 0.1 \times (10 \times 100 \times \cos 100t + 3 \times 300 \times \cos (300t + \pi/3) + 2 \times 500 \times \cos(500t - \pi/6)) \\
 &= 10 + 141.4 \sin (100t + \pi/4) + 94.8 \sin (300t + \pi/3 + 71.56^\circ) + 101 \sin (500t - \pi/6)
 \end{aligned}$$

[2 marks]

- (c) Figure Q5c shows a circuit which is switched at time $t = 0$. Draw the corresponding Laplace transformed circuit



[1 mark]

Write an expression for the current $I_s(s)$ in the Laplace domain

$$I(s) = \frac{200 \times \frac{250}{s^2 + 250^2}}{50 + 0.08s + \frac{0.12s \times \frac{10^4}{s}}{0.12s + \frac{10^4}{s}}} = \frac{200 \times 250 \times (0.12s^2 + 10^4)}{(s^2 + 250^2) \left[(50 + 0.08s)(0.12s^2 + 10^4) + 0.12 \times 10^4 \right]}$$

[2 marks]

Hence show how the corresponding current $i_s(t)$ may be determined.

$$I(s) = \frac{As + B}{s^2 + 250^2} + \frac{C}{s + \alpha} + \frac{D}{s + \beta} + \frac{E}{s + \gamma}$$

$$i(t) = A \cos 250t + \frac{B}{250} \sin 250t + C \cdot e^{-\alpha t} + D \cdot e^{-\beta t} + E \cdot e^{-\gamma t}$$

[2 marks]

[END OF QUESTION PAPER]